

Ψ_s – estimated value of the magnetic flux of the motor,
 m_e – estimated value of the electromagnetic torque of the motor,
 M – low power induction motor,
 T – tachogenerator.

What is characteristic of this system is the fact that the sinus and cosinus of the γ_s angle are calculated based on the measurements of phase currents and voltages and based on the mathematical model of a slip-ring shaded-pole motor (the mathematical model parameters of the this motor were calculated earlier by a different evolutionary algorithm [7, 8, 10]). The process of parametric optimization [11] was conducted with the use of an evolutionary algorithm. The calculations of parameters required for the controllers were made [3, 4, 5]. The results of these calculations can be seen in Table 1.

Table 1.
Calculated settings of PI controllers of the DFOC system, with the use of an evolutionary algorithm

$K_{p,1}$	$K_{p,2}$	$K_{p,3}$	$K_{p,4}$	$K_{p,5}$	$T_{p,1}$	$T_{p,2}$	$T_{p,3}$	$T_{p,4}$	$T_{p,5}$	F_i [rotations/min]
2.00	5.00	5.00	9.00	15.0	0.40	0.30	2.00	0.10	0.10	$2.27 \cdot 10^5$

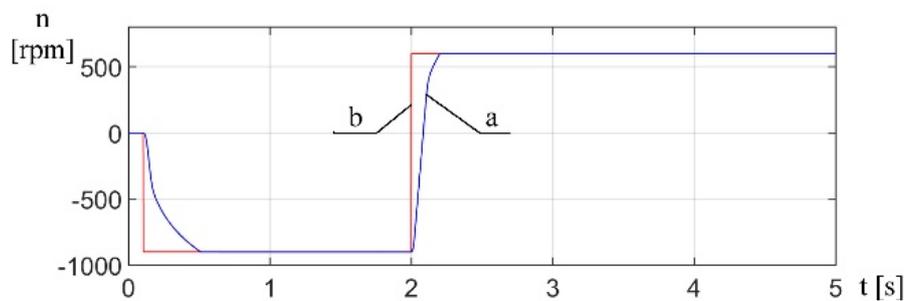


Fig. 2. Waveforms of rotational speed: the set process (b) and the process executed by the control system based on the calculated parameters of PI controllers (a)

So far the calculations have been focused on proper and sub-optimal calculation of settings of PI controllers for different versions of the FOC system and for different motors (squirrel-cage, slip-ring shaded-pole motors) [1, 2, 3, 4, 5, 9, 10]. The impact of different evolutionary algorithm parameters on the obtained evolutionary results was investigated. Further in the work the authors tested the impact of sine-wave disturbances introduced to input signals of a control system on output waveforms (rotational speed waveform and electromagnetic torque waveform). These tests were based on simulations carried out with the use of MATLAB/Simulink.

Symbols used in the table:

- $K_{p,1}$ – boosting of the current controller in the control loop of the magnetic flux,
- $K_{p,2}$ – boosting of the current controller in the control loop of the speed controller,
- $K_{p,3}$ – boosting of the magnetic flux controller,
- $K_{p,4}$ – boosting of the electromagnetic torque controller,
- $K_{p,5}$ – boosting of the rotational speed controller,
- $T_{p,1}; T_{p,2}; T_{p,3}; T_{p,4}; T_{p,5}$ – coefficients dependent on integral action times of the controllers in the control loop, as above.
- F – quality criterion which is a total of modules of differences in the value of rotational speed generated on the basis of current settings of controllers and the value of rotational speed set to the control system in discrete moments of time (simulation time step – 0.001 s, simulation time – 5 s).

The processes set for the described control system were the following:

- step change of rotational speed which was to be executed by the DFOC system,
- step change of load torque after the rotational speed sets in.

Ideal waveforms of rotational speed (Fig. 2): the set process (b) and the process executed by the control system (a).

3. TESTING DFOC RESISTANCE TO SINE-WAVE DISTURBANCES

A block diagram of each linear controller of the tested control system can be seen in Fig. 3. Here an extra input was assumed through which a sine-wave disturbing signal entered. This signal was then added to the input signal of a linear controller. The impact of such disturbances on the system dynamics [6] (Fig. 3) was tested with the use of a known-amplitude sine and frequency as an additive component of the controller input signal.

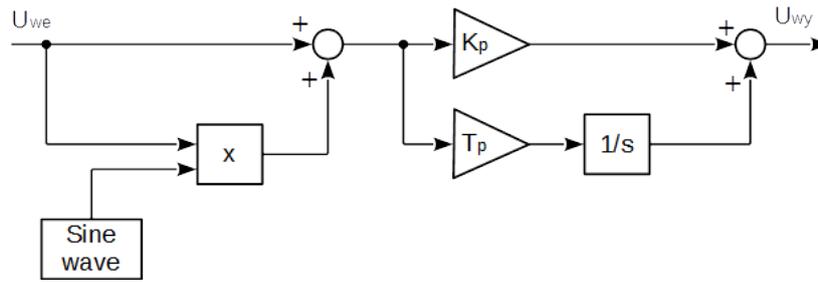


Fig. 3. Diagram of the controller with sine-wave disturbances in the input signal:
 U_{we} – input signal of the controller; U_{wy} – output signal of the controller; K_p – controller boosting;
 T_p – coefficient dependent on integral action times

The system was tested in the following manner: an additive sine-wave disturbing signal was introduced to all controllers simultaneously, then rotational speed waveform and electromagnetic torque waveform were observed,

The output measure was the mean absolute percentage error (MAPE) expressed by a commonly known formula [3, 5]:

$$MAPE = \frac{1}{n} \sum_{i=1}^n \left| \frac{x_{zi} - x_i}{x_{zi}} \right| 100\% \quad (1)$$

where:

- $MAPE$ – value of mean absolute percentage error,
- n – number of moments of time in which the values of errors were calculated ($n = 5000$, $\Delta t = 0.001$ s),
- x_{zi} – set value of rotational speed measured in the successive i -th moment of time,
- x_i – output value of the FOC control system in the successive moment of time i .

The calculation results are presented in Table 2.

Selected waveforms of rotational speed to be conducted by the DFOC control system with a slip-ring shaded-pole motor can be seen in Fig. 5-7.

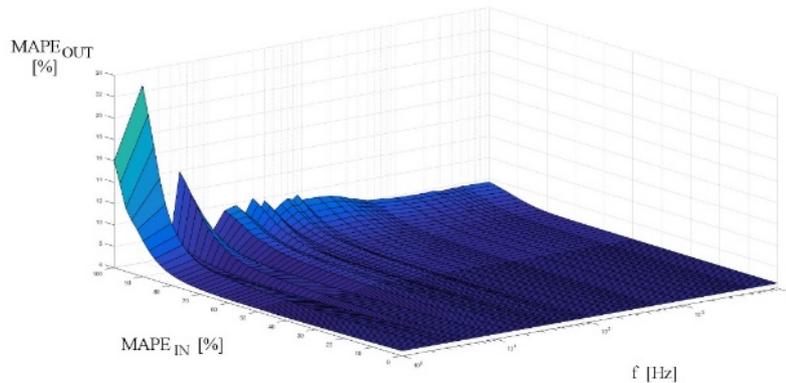


Fig. 4. Dependency of the parameters of an additive sine-wave disturbing signal (frequency f and the contents of this disturbing signal in the basic signal $MAPE_{IN}$) on the level of disturbances in the output signal $MAPE_{OUT}$

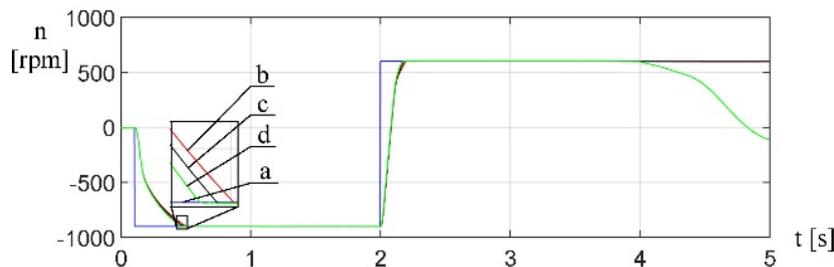


Fig. 5. Response of a control system with introduced additive disturbing signals for 1Hz on the level $b - 10\%$; $c - 50\%$; $d - 100\%$ for the set step of rotational speed and load step in the 4th second of simulation – a

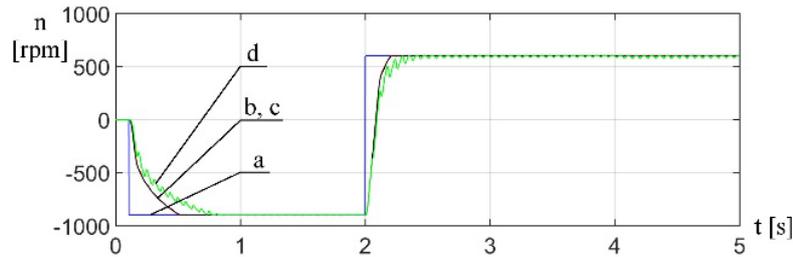


Fig. 6. Response of a control system with introduced additive disturbing signals for 1Hz on the level b – 10%; c – 50%; d – 100% for the set step of rotational speed and load step in the 4th second of simulation – a

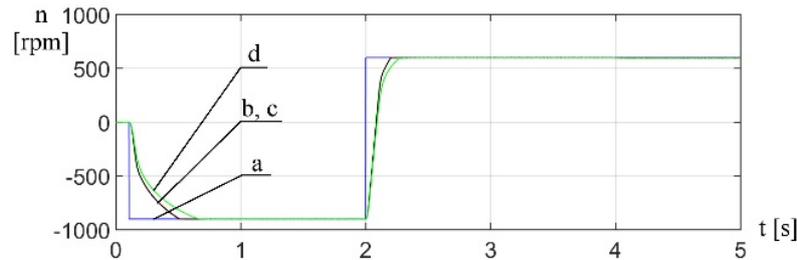


Fig. 7. Response of a control system with introduced additive disturbing signals for 1kHz on the level b – 10%; c – 50%; d – 100% for the set step of rotational speed and load step in the 4th second of simulation – a

4. CONCLUSIONS

A non-zero average value of the absolute percentage error of the output quantity of the system with no introduced disturbances results from the fact (Fig. 2) that the set value of the rotational speed is not equal to the value of the rotational speed conducted by the control system. In the case of a control system with no disturbing signals introduced, there is a minimal overshoot and the control time does not exceed 0.5 s. In the 4th second the system was loaded with an external torque and the system responded with immediate stabilization of rotations. After the disturbances were introduced to internal signals of the control system, the system behaved similarly up to the disturbance level of 50%. Above this level of the MAPE_{IN} value, the system was unstable and there were visible vibrations of the rotational speed. This effect was characteristic of low frequency values, up to 1 kHz. The lower was the frequency of additive disturbing signals, the more unstable was the system (Fig. 4, 5). Therefore it is possible to deduct the following: the DFOC system tolerates sine-wave additive disturbing signals up to the value of their 50% content and is stable. Once this boundary is crossed, the machines will wear out more quickly and their operations will be unexpected.

Bibliography

- Głowacz A., Głowacz Z.: Diagnostics of induction motor based on analysis of acoustic signals with application of FFT and classifier based on words. Archives of Metallurgy and Materials 55(3), 707 – 712, 2010
- Hudy W.: Analysis of parametric optimization of field-oriented control of 3-Phase induction motor with using evolutionary algorithm, Transactions of the VŠB – Technical University of Ostrava, Mechanical Series 2(59), art. No. 1959, 2013

- Hudy W., Jaracz K.: Wpływ rodzaju wskaźnika jakości na sygnał prędkości obrotowej w układzie DFOC przy wprowadzonych sygnałach zakłócających (Impact of quality coefficient type on rotational speed signal in DFOC system with introduced disturbing signals), 40th Conference ATI'2015 Automation, Telecommunications, Information Technology, Szczyrk, 245-252, 2015
- Hudy W., Jaracz K.: Evolutionary operators impact on results of evolutionary parametric optimization of FOC system with induction motor, 39th Conference ATI'2013 Automation, Telecommunications, Information Technology, Zakopane, 2013
- Hudy W., Noga H.: Influence of various types of interference of entry signals' of regulators type pi in field oriented control system with induction motor on initial rotational speed. 2014 International Conference on Energy Materials and Environment Engineering ICEMEE 2014, Guangzhou, China
- Jaracz K.: Rozszerzone modele wrażliwości maszyn prądu stałego przy zakłóceniach niezdeterminowanych (Extended models of DC machines sensitivity with undetermined disturbances), Scientific Publishing WSP, Kraków 1998
- Kaźmierkowski M.P.: Porównanie metody sterowania polowozorientowanego z metodą bezpośredniej regulacji momentu silnika klatkowego (Comparison of field-oriented control method with the method of direct control of the squirrel-cage motor torque). Electro-technical Review 4(98), Warszawa, 1998
- Leonard W.: Control of Electrical Drives. Springer Verlag, Berlin 1985
- Miksiewicz R.: Maszyny elektryczne (Electrical machines). Silesian University of Technology Press, 2000
- Orłowska-Kowalska T.: Control systems of the induction motors without sensors. Wrocław University of Technology Press, Wrocław, 2003
- Vítečková, M., Víteček, A.: Vybrané metody seřizování regulátorů. VŠB-TU Ostrava, 2011
- Walek, B., Farana, R.: A tool for searching in information systems under uncertainty. AIP Conference Proceedings 1(1738), 2016

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